

clamps for the luma and composite.

powered by a single +5V supply.



### **Complete Audio/Video Backend Solution**

#### **Features**

- ♦ Integrated Video Reconstruction Filters—6MHz Lowpass Filter Supports NTSC, PAL, or DVB per ITU-601
- ♦ Integrated Video and Audio Amplifiers
- ♦ Mono Audio and CVBS Output to Drive External Modulator
- ♦ +5V (Video) and +9V to +12V (Audio) Single-**Supply Operation**
- ♦ Differential/Single-Ended Audio Inputs

- ♦ Integrated Video Input Clamps and Biasing

- **♦ 24-Pin TSSOP Package**

### **Ordering Information**

PART	TEMP RANGE	PIN-PACKAGE
MAX4079CUG+	0°C to +70°C	24 TSSOP

<sup>+</sup>Denotes lead(Pb)-free package.

**General Description** 

The MAX4079 filters and buffers video (NTSC/PAL/DVB)

and stereo audio signals from the MPEG decoder of a

cable/satellite receiver, VCR/DVD player, or a TV to an

external load. The MAX4079 has luma-chroma (Y-C)

and composite (CVBS) video inputs with one Y-C and two CVBS outputs. All video inputs are AC-coupled with

internal DC biasing on the chroma input and active

The MAX4079 video reconstruction filters have a 6MHz cutoff frequency and 50dB attenuation at 27MHz. The

filters are matched with flat group delay for standard-

definition video. The video gain is fixed at +6dB to drive

a 75 $\Omega$  back-terminated load (150 $\Omega$ ) to unity gain. The video outputs can be either DC- or AC-coupled and are

The MAX4079 audio amplifiers have differential inputs for optimum performance, but can be used with single-

ended sources with external biasing. The audio channels have a fixed gain of +6dB and deliver 2.6VRMS output with a differential input of ±1.85V. The audio amplifiers operate from a +9V to +12V single supply

and feature an internal bias generator. An on-chip mixer also provides a mono output, with +3dB gain,

The MAX4079 is available in 24-pin TSSOP package. and is fully specified over the 0°C to +70°C commercial

temperature range. The MAX4079 Evaluation Kit is

derived from the left and right audio channels.

Satellite Receivers

available to help speed designs.

Cable Receivers

Home Theater Systems

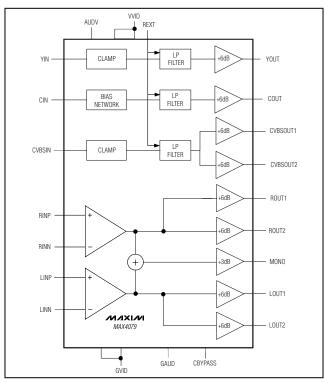
**AV Receivers** 

**Televisions** 

### **Applications**

**DVD Players** 

### Functional Diagram



Pin Configuration appears at end of data sheet.

#### **ABSOLUTE MAXIMUM RATINGS**

GVID to GAUD0.1V to +0.1 VVID to GVID0.3V to +6	
AUDV to GAUD0.3V to +14	
LINP, LINN, RINP, RINN, CBYPASS to GAUD0.3V to +6	δV
LOUT1, LOUT2, ROUT1, ROUT2,	
MONO to GAUD0.3V to lower of (+9V and AUDV + 0.3)	V)
YIN, CIN, CVBSIN, REXT to GVID0.3V to (VVID + 0.3)	V)
YOUT, COUT, CVBSOUT1, CVBSOUT2 to	
GVID0.3V to (VVID + 0.3)	V)
Video Output Short-Circuit Duration to GVID or	

Audio Output Short-Circuit Duration to GAUD	
AUDV	Continuous
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
24-Pin TSSOP (derate 12.2mW/°C above +7	0°C)975.6mW
Operating Temperature Range	0°C to +70°C
Storage Temperature Range	
Junction Temperature	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### DC ELECTRICAL CHARACTERISTICS

 $(V_{VVID} = +5V, V_{AUDV} = +12V, V_{GVID} = V_{GAUD} = 0V, R_{LOAD\_VID} = 150\Omega \ to \ GVID, R_{REXT} = 10k\Omega \ \pm 1\%, C_{CBYPASS} = 1\mu F, T_A = 0^{\circ}C \ to +70^{\circ}C, unless otherwise noted. Typical values are at T_A = +25^{\circ}C.) (Note 1)$ 

PARAMETER	SYMBOL	CO	NDITIONS	MIN	TYP	MAX	UNITS
SUPPLIES	•						
Audio Supply Voltage Range	AUDV			8.5		12.6	V
Video Supply Voltage Range	VVID			4.75	5.0	5.25	V
Video Quiescent Supply Current	Iccv	V <sub>VVID</sub> = 5.25V, no AC-coupled to gro	load, all video inputs ound		60	100	mA
Audio Quiescent Supply Current	ICCA	V <sub>AUDV</sub> = 12.6V, no biased at 2.5V	V <sub>AUDV</sub> = 12.6V, no load, audio inputs			15	mA
Thermal Shutdown	T <sub>SD</sub>	Rising die tempera	ature		+150		°C
Thermal-Shutdown Hysteresis	T <sub>SD,HYS</sub>				25		°C
VIDEO							
Voltage Gain	Av,vid	V <sub>IN</sub> = 1V <sub>P-P</sub> , all vic	V <sub>IN</sub> = 1V <sub>P-P</sub> , all video inputs, no load		6	6.2	dB
Gain Matching	ΔA <sub>V,VID</sub>	V <sub>IN</sub> = 1V <sub>P-P</sub> , all vic	V <sub>IN</sub> = 1V <sub>P-P</sub> , all video inputs, no load			+0.4	dB
	V <sub>IN,VID</sub>	YIN, CVBSIN		0		1.2	V <sub>P-P</sub>
Input Voltage Swing		CIN		0		0.9	VP-P
Clamp Voltage	VCLMP	CVBSOUT_ and Y	OUT, no signal, no load		1.0		V
Chroma Bias	V <sub>BIAS</sub>	COUT, no signal, r	no load		2.1		V
Droop	D	(Note 2)				2	%
REXT Reference Voltage	V <sub>REXT</sub>			0.85	1.00	1.15	V
Innut Decistors	D	CVBSIN or YIN			2.3		МΩ
Input Resistance	R <sub>IN,VID</sub>	CIN			10		kΩ
Input Clamping Current	ICLMP	CVBSIN or YIN input, V <sub>IN</sub> = 3.5V		1	2.5	4	μΑ
Output Voltage Swing	\/o.u.=.v:=	CVBSOUT_, YOUT	Ţ	2.4			\/= =
Output Voltage Swing	ige Swing Vout,vid (		COUT				V <sub>P-P</sub>
Short-Circuit Current	ISC,VID	Video output short	ed to VVID or GVID		50		mA
Dower Cupply Dejection Datio	DCDD: #P	4.75V ≤ V <sub>VVID</sub> ≤	YOUT/COUT		48		٩D
Power-Supply Rejection Ratio	PSRR <sub>VID</sub>	5.25V	CVBSOUT_		48		dB

#### DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{VVID} = +5V, \ V_{AUDV} = +12V, \ V_{GVID} = V_{GAUD} = 0V, \ R_{LOAD\_VID} = 150\Omega \ to \ GVID, \ R_{REXT} = 10k\Omega \ \pm 1\%, \ C_{CBYPASS} = 1\mu F, \ T_A = 0^{\circ}C \ to +70^{\circ}C, \ unless \ otherwise \ noted. \ Typical values are at \ T_A = +25^{\circ}C.) \ (Note 1)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
AUDIO						
Voltage Gain	Av,aud	1.414Vp-p differential input	5.8	6	6.2	dB
Mono Voltage Gain	Av,mono	1.414Vp-p differential input, LIN = RIN	2.8	3	3.2	dB
Gain Matching Between Channels	ΔA <sub>V,AUD</sub>	1.414V <sub>P-P</sub> differential input	-0.4		+0.4	dB
Input Voltage Range	V <sub>IN,AUD</sub>	Inferred from CMRR test	0.3		5.2	V
Differential Input Voltage Range	V <sub>IN</sub> ,AUD DIF	Inferred from output voltage swing	-1.85		+1.85	V
Input Current	I <sub>IN,AUD</sub>				2	μΑ
Output Voltage Swing	Vout, aud	Input overdriven, 10kΩ load to 4.15V	7.4			V <sub>P-P</sub>
Short-Circuit Current	ISC,AUD			15		mA
Power-Supply Rejection Ratio	PSRR <sub>AUD</sub>	8.5V ≤ V <sub>AUDV</sub> ≤ 12.6V	70			dB
Common-Mode Rejection Ratio	CMRR <sub>AUD</sub>	$0.3V \le V_{CM} \le 5.2V$	50	60		dB

#### **AC ELECTRICAL CHARACTERISTICS**

 $(V_{VVID} = +5V, \ V_{AUDV} = +12V, \ V_{GVID} = V_{GAUD} = 0V, \ R_{IN\_VIDEO} = 75\Omega \ to \ GVID, \ C_{IN\_VIDEO} = 0.1 \mu F, \ R_{LOAD\_VID} = 150\Omega \ to \ GVID, \ C_{OUT\_AUDIO} = 10 \mu F, \ R_{LOAD\_AUD} = 10 k\Omega \ \pm 1\% \ to \ GAUD, \ R_{REXT} = 10 k\Omega, \ C_{CBYPASS} = 1 \mu F, \ T_A = 0^{\circ}C \ to \ +70^{\circ}C, \ unless \ otherwise noted. \ Typical values are at T_A = +25^{\circ}C.) \ (Note 1)$ 

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
VIDEO							
		CVBSOUT1 = CVBSOUT2 = YOUT = COUT = 2VP-P;	f = 4MHz	-0.5		+0.5	
Filter Attenuation	AVIDEO	$R_L = 150\Omega$ to ground,	f = 7MHz		3		dB
		attenuation is referred to 100kHz	f = 27MHz	40	50		
Slew Rate	SR	V <sub>OUT</sub> = 2V <sub>P-P</sub>			30		V/µs
Differential Gain	DG	CVBSOUT_, YOUT, COUT, 5 modulated staircase		0.5		%	
Differential Phase	DP	CVBSOUT_, YOUT, COUT, 5 modulated staircase		0.9		degrees	
Power-Supply Rejection Ratio	PSRRvin	f = 100kHz, 0.5V <sub>P-P</sub>	YOUT/COUT		48	dD	dB
Fower-Supply nejection hatto	FOUNT	T = 100kHz, 0.5VP-P	CVBSOUT_		44		иь
Peak Signal to RMS Noise	SNR <sub>VID</sub>	CVBSOUT_, YOUT, COUT, V	IN = 1V <sub>P-P</sub>		65		dB
Group Delay Deviation	GD	CVBSOUT_, YOUT, COUT, f <sub>IN</sub> = 0.1MHz to 4.5MHz			25		ns
Output Impedance	Z <sub>OUT,VID</sub>	f = 3.58MHz			0.5		Ω
Capacitive Load	C <sub>L,VID</sub>	No sustained oscillations			35		pF
Video Crosstalk	X <sub>TALK,VID</sub>	f = 3.58MHz, 1V <sub>P-P</sub> input, between any two active inputs			-63		٩D
Audio/Video Crosstalk	XTALK,VD/AD	f = 15kHz, 1V <sub>P-P</sub> input, betwee active audio or video inputs	een any two		-76		dB

#### **AC ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{VVID} = +5V, \ V_{AUDV} = +12V, \ V_{GVID} = V_{GAUD} = 0V, \ R_{IN\_VIDEO} = 75\Omega \ to \ GVID, \ C_{IN\_VIDEO} = 0.1\mu F, \ R_{LOAD\_VID} = 150\Omega \ to \ GVID, \ C_{OUT\_AUDIO} = 10\mu F, \ R_{LOAD\_AUD} = 10k\Omega \ \pm 1\% \ to \ GAUD, \ R_{REXT} = 10k\Omega, \ C_{CBYPASS} = 1\mu F, \ T_A = 0^{\circ}C \ to \ \pm 70^{\circ}C, \ unless \ otherwise noted. \ Typical values are at T_A = \pm 25^{\circ}C.) \ (Note 1)$ 

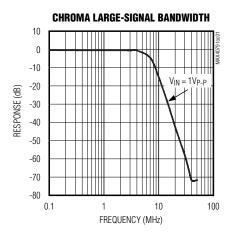
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
AUDIO						
Gain Flatness	ΔA <sub>AUD</sub>	LOUT_, ROUT_, $f_{IN}$ = 20Hz to 20kHz, 0.5V <sub>RMS</sub> input		0.01		dB
	Zi (AOD	MONO, $f_{IN} = 20Hz$ to $20kHz$ , $0.5V_{RMS}$ input				ub
Signal-to-Noise Ratio	SNR <sub>AUD</sub>	f <sub>IN</sub> = 1.0kHz, 0.5V <sub>RMS</sub> , CCIR weighing highpass filter at 20Hz, lowpass filter at 20kHz		85		dB
Total Harmonic Distortion	THD+N	$f_{IN} = 1.0kHz, 0.5V_{RMS}$		0.005		%
Plus Noise	THD+N	$f_{IN} = 1.0kHz, 1V_{RMS}$		0.003		%
Output Impedance	Z <sub>O,AUD</sub>	f = 1kHz		0.2		Ω
Power-Supply Rejection Ratio	PSRR <sub>AUD,</sub> AC	f = 1kHz, V <sub>RIPPLE</sub> = 200mV <sub>P-P</sub>		60		dB
Crosstalk	X <sub>TLK</sub> ,AUD	$f = 1kHz, 0.5V_{RMS}$ input		70		dB
Capacitive Load	C <sub>L,AUD</sub>	No sustained oscillations		200	•	pF

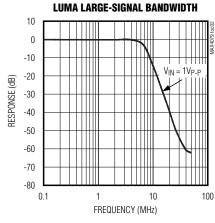
**Note 1:** All devices are 100% production tested at  $T_A = +25^{\circ}C$ . Specifications over temperature limits are guaranteed by design.

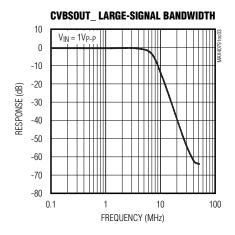
Note 2: Droop is defined as the percentage change in the DC level from the start to the end of a video line. Inferred from input clamping current with a 0.1μF coupling capacitor.

### Typical Operating Characteristics

 $(V_{VVID} = +5V, \ V_{AUDV} = +12V, \ V_{GVID} = V_{GAUD} = 0V, \ R_{IN\_VIDEO} = 75\Omega \ to \ GVID, \ C_{IN\_VIDEO} = 0.1 \mu F, \ R_{LOAD\_VID} = 150\Omega \ to \ GVID, \ C_{OUT\_AUDIO} = 10 \mu F, \ R_{LOAD\_AUD} = 10 k\Omega \ to \ G_{AUD}, \ R_{REXT} = 10 k\Omega, \ C_{BYPASS} = 1 \mu F, \ T_A = +25 ^{\circ}C, \ unless \ otherwise \ noted.)$ 

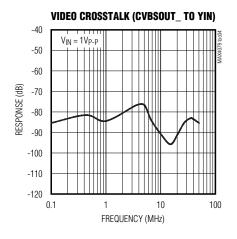


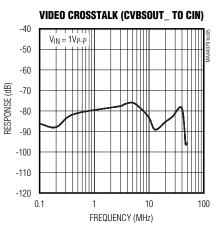


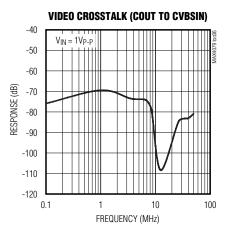


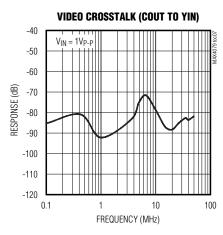
### Typical Operating Characteristics (continued)

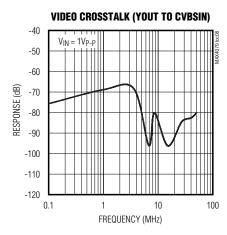
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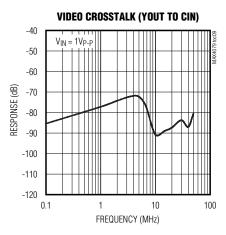


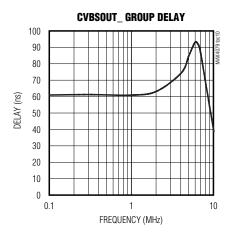


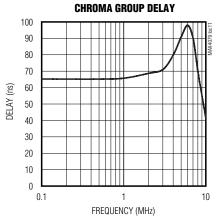


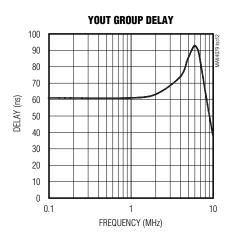






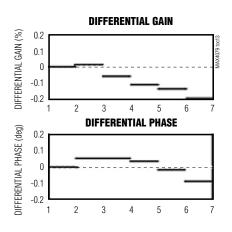


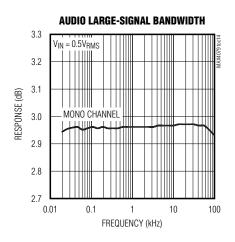


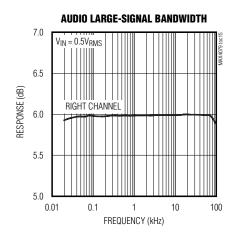


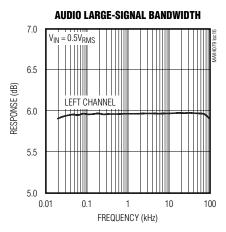
#### Typical Operating Characteristics (continued)

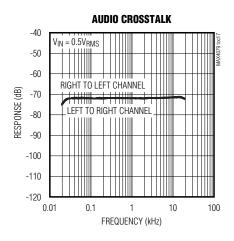
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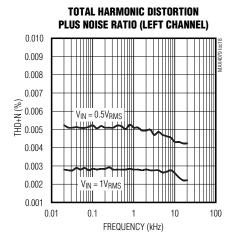






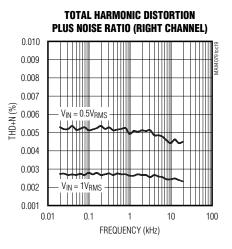


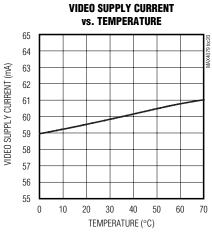


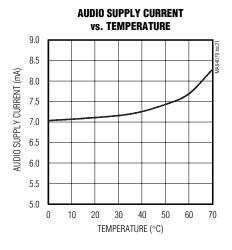


### Typical Operating Characteristics (continued)

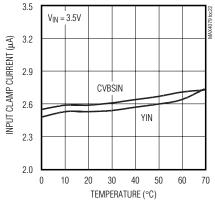
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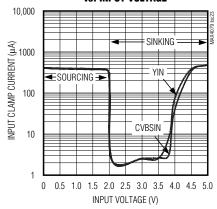




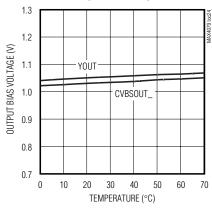
## INPUT CLAMP CURRENT vs. TEMPERATURE



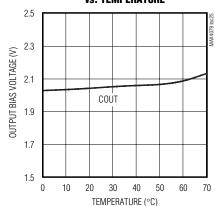
## INPUT CLAMP CURRENT vs. INPUT VOLTAGE



## OUTPUT BIAS VOLTAGE vs. TEMPERATURE



## OUTPUT BIAS VOLTAGE vs. TEMPERATURE



#### **Pin Description**

PIN	NAME	FUNCTION
1	REXT	External Filter Resistor. Bypass with a $10k\Omega \pm 1\%$ resistor and a $0.1\mu F$ capacitor to GVID.
2, 24	VVID	Video Power-Supply Input
3	CVBSIN	Composite Video Input
4, 21	GVID	Video Ground
5	YIN	Luma Input
6	CBYPASS	Audio LDO Regulator Bypass Capacitor. Bypass with a1µF capacitor to GAUD.
7	CIN	Chroma Input
8	AUDV	Audio Power-Supply Input
9	LINP	Left-Channel Audio Positive Input
10	LINN	Left-Channel Audio Negative Input
11	RINN	Right-Channel Audio Negative Input
12	RINP	Right-Channel Audio Positive Input
13	ROUT2	Right-Channel Audio Output 2
14	ROUT1	Right-Channel Audio Output 1
15	GAUD	Audio Ground
16	MONO	Mono Audio Output
17	LOUT2	Left-Channel Audio Output 2
18	LOUT1	Left-Channel Audio Output 1
19	COUT	Chroma Video Output
20	YOUT	Luma Video Output
22	CVBSOUT1	Composite Video Output 1
23	CVBSOUT2	Composite Video Output 2

### **Detailed Description**

The MAX4079 filters and buffers video (NTSC/PAL/DVB) and stereo audio signals from the MPEG decoder of a cable/satellite receiver, VCR/DVD player, or a TV to an external load. The MAX4079 has luma-chroma (Y-C) and composite (CVBS) video inputs with one Y-C and two CVBS outputs. All video inputs are AC-coupled with internal DC biasing on the chroma input and active clamps for the luma and composite.

The MAX4079 video reconstruction filters have a 6MHz cutoff frequency and 50dB attenuation at 27MHz. The filters are matched, with flat group delay for standard-definition video. The video gain is fixed at +6dB to drive a 75 $\Omega$  back-terminated load (150 $\Omega$ ) to unity gain. The video outputs can be either DC- or AC-coupled and are powered by a single +5V supply.

The MAX4079 audio amplifiers have differential inputs for optimum performance, but can be used with single-ended sources with external biasing. The audio channels have a fixed gain of  $\pm 6dB$  and deliver  $2.6V_{RMS}$  output with a differential input of  $\pm 1.85V$ . The audio

amplifiers operate from a +9V to +12V single supply and feature an internal bias generator. An on-chip mixer also provides a mono output, with +3dB gain, derived from the left and right audio channels.

#### **Video**

The video section of the MAX4079 implements DC restore/biasing, amplification, and reconstruction filtering for the Y-C and CVBS input signals. All of the video inputs are AC-coupled. DC restore is performed using a sync tip clamp for both luma and composite video channels. The chroma DC level input is biased at the midlevel of the signal.

All video channels have a fixed gain of +6dB. The DC level at the video outputs is controlled so that coupling capacitors are not required.

All composite and luma video outputs are capable of driving 2.4VP-P, and the chroma output is capable of driving 1.8VP-P into 150 $\Omega$  resistive load to ground. Up to 35pF of load capacitance can be tolerated at each video output without stability or slew-rate issues.

\_\_ /N/XI/N

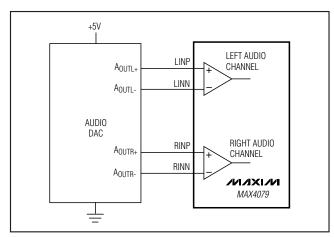


Figure 1. Differential Audio Inputs

All video inputs are stable with up to  $150\Omega$  source resistance. For higher values, consult Maxim applications.

#### Video Reconstruction Filter

The MAX4079 reconstruction filters are 4th-order Butterworth filters that provide a cutoff frequency of 6MHz and flat group delay response up to 4.5MHz. The stopband offers 26dB of attenuation at 13.5MHz and 50dB at 27MHz.

#### **Audio**

The audio section of the MAX4079 is a stereo amplifier with one differential input and two single-ended outputs for each channel (left and right). A mono output is provided by summing the two channels of the stereo signal together. The stereo channels have a +6dB typical gain, while the mono has a +3dB gain.

The audio inputs can be DC-coupled, eliminating space-consuming coupling capacitors. Each of the five outputs can deliver 2.6VRMs into an AC-coupled  $10k\Omega$  load.

### Applications Information

#### **Audio DAC Interfacing**

#### Differential Audio DAC

The MAX4079 accepts differential audio signals. Figure 1 shows a typical configuration for connecting the device to an audio DAC with differential outputs. Figure 2 shows the reconstruction filters that can be used for the differential audio inputs. Carefully select resistors and capacitors to attenuate out-of-band noise and mini-

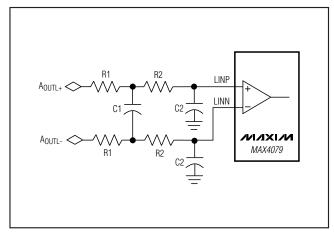


Figure 2. Filtering Differential Audio Inputs

mize the effect on the gain. The common-mode voltage for these signals is typically 2.5V.

#### Single-Ended Audio DAC

The MAX4079 can also be configured for single-ended inputs. Figure 3 shows how to connect an audio DAC with single-ended outputs to the MAX4079. Figure 4 shows how reconstruction filters can be used for single-ended audio inputs. Choose these values to minimize the effect on gain.

If the single-ended audio DAC does not include a common-mode voltage output, create a bias point with well-matched resistors and couple the audio signal to the positive differential input (see Figure 5). The bias point can also be created using a resistor-divider network from the video supply voltage. Note that the tolerance of the resistors will affect the common-mode and power-supply rejection ratios. Tighter tolerances improve the performance of CMRR and PSRR, e.g., 1% resistors will not give any better than 40dB of CMRR and PSRR, whereas 0.1% resistors could improve the number to 60dB.

#### **Power Supplies and Bypassing**

The MAX4079 features single +5V (video) and +12V (audio) supply operation, and requires no negative supply. Connect the VVID pins together and bypass to GVID with  $0.01\mu\text{F}$ ,  $0.1\mu\text{F}$ , and  $4.7\mu\text{F}$  capacitors in parallel. Bypass the AUDV to GAUD with  $0.1\mu\text{F}$ ,  $1\mu\text{F}$ , and  $47\mu\text{F}$  capacitors in parallel. Bypass CBYPASS to GAUD with a  $1\mu\text{F}$  capacitor (see the *Typical Operating Circuit*).

#### **Layout and Grounding**

For optimal performance, stitch ground vias between the narrow adjacent signal traces to minimize crosstalk. Avoid running video traces parallel to high-speed data lines. The MAX4079 provides separate ground connections

for video and audio supplies. For best performance use separate ground planes for each of the ground returns, and connect all ground planes together at a single point. Refer to the MAX4079 Evaluation Kit for a proven circuit board layout example.

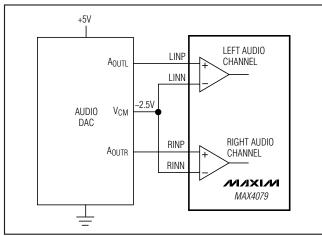


Figure 3. Single-Ended Audio Inputs

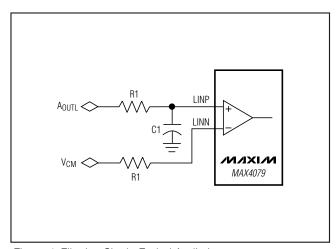


Figure 4. Filtering Single-Ended Audio Inputs

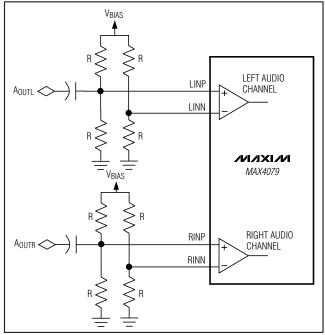


Figure 5. Biasing Single-Ended Audio Inputs

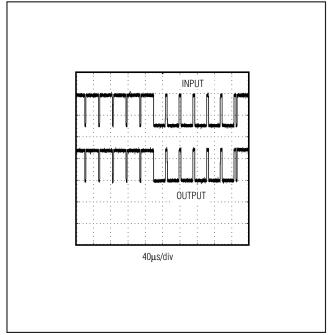
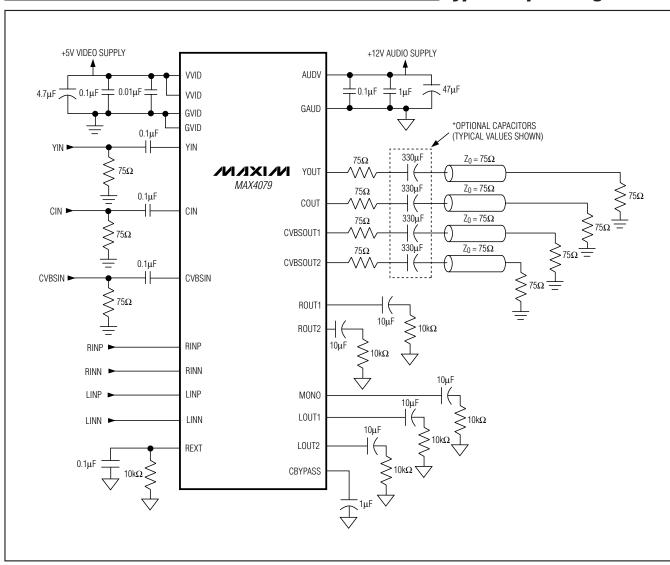


Figure 6. Vertical Sync Interval

### **Typical Operating Circuit**



#### **Pin Configuration** TOP VIEW REXT 1 24 VVID VVID 2 23 CVBSOUT2 22 CVBSOUT1 CVBSIN 3 21 GVID GVID 4 MIXKM MAX4079 YIN 5 20 YOUT 19 COUT CBYPASS 6 CIN 7 18 LOUT1 AUDV 8 17 LOUT2 16 MONO LINP 9 LINN 10 15 GAUD RINN 11 14 ROUT1 13 ROUT2 RINP 12 **TSSOP**

\_\_Chip Information

PROCESS: BiCMOS

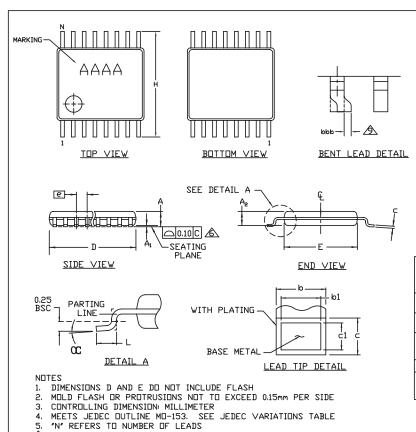
TSSOP4.40mm.EPS

### **Complete Audio/Video Backend Solution**

#### **Package Information**

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
24 TSSOP	U24+1	<u>21-0066</u>



S M B D L	C	MMON D	SMENSIONS		
M B	MILLIM	ETERS	INC	<del>I</del> ES	
٦.	MIN.	MAX.	MIN.	MAX.	
Α		1.10		.043	
A <sub>1</sub>	0.05	0.15	.002	.006	
Aε	0.85	0.95	.033	.037	
b	0.19	0.30	.007	.012	
b <sub>1</sub>	0.19	0.25	.007	.010	
C	0.09	0,20	.004	.008	
C1	0.09	0.14	.004	.006	
D	SEE VAR	ZNDITAIS	SEE VAR	ZNDITAIS	
Ε	4.30	4.50	.169	.177	
е	0.65	BSC	.026	BSC	
Н	6.25	6.55	.246	.258	
L	0.50	0.70	.020	.028	
N	SEE VAR	ZNDITAIS	SEE VARIATIONS		
Œ	0°	8°	0°	8°	
bbb	0.10	MAX			

					VARIA	ZNDITA	
JEDEC	١		PKG.	MILLIM	IETERS	INCH	<del>I</del> ES
MD-153	N		CODES	MIN.	MAX.	MIN.	MAX.
AB-1	14	D	U14-1; U14-2	4.90	5.10	.193	.201
AB	16	D	U16-1; U16-2	4.90	5.10	.193	.201
AC	20	D	U20-2; U20M-2 U20-3	6.40	6.60	.252	.260
AD	24	D	U24-1	7.70	7.90	.303	.311
AE	28	D	U28-1; U28-2; U28-3	9.60	9.80	.378	.386

PACKAGE DUTLINE, TSSOP 4.40mm BODY

APPROVAL	DOCUMENT CONTROL NO.	REV.	1 /
	21-0066	J	1/1

٠.	11 11	LI LIVO 10 140	III IDEIN	ш.	LLHDS
<b>∕6</b> \.	LEAD	COPLANARIT'	Y 0.10	MM	MAX.

- 7. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY

  8. MARKING IS FOR PACKAGE DRIENTATION REFERENCE ONLY

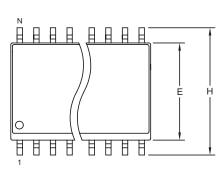
  \$\frac{1}{2}\$ BENT LEAD 0.10 MM MAX.

- 10. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PBFREE (+) PKG. CODES.

-DRAWING NOT TO SCALE-

### **Package Information (continued)**

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

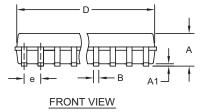


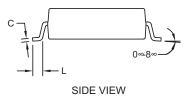
		INC	HES	MILLIM	ETERS
	DIM	MIN	MAX	MIN	MAX
	Α	0.093	0.104	2.35	2.65
	A1	0.004	0.012	0.10	0.30
	В	0.014	0.019	0.35	0.49
	С	0.009	0.013	0.23	0.32
	е	0.0	)50	1.2	27
	Е	0.291	0.299	7.40	7.60
ĺ	Н	0.394	0.419	10.00	10.65
	L	0.016	0.050	0.40	1.27

#### **TOP VIEW**

#### VARIATIONS:

		INCHES		MILLIM			
	DIM	MIN	MAX	MIN	MAX	Ν	MS013
	D	0.398	0.413	10.10	10.50	16	AA
	D	0.447	0.463	11.35	11.75	18	AB
	D	0.496	0.512	12.60	13.00	20	AC
	D	0.598	0.614	15.20	15.60	24	AD
	D	0.697	0.713	17.70	18.10	28	AE





#### NOTES:

- 1. D&E DO NOT INCLUDE MOLD FLASH.
- 2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED 0.15mm (.006").
- 3. LEADS TO BE COPLANAR WITHIN 0.10mm (.004").
  4. CONTROLLING DIMENSION: MILLIMETERS.
- 5. MEETS JEDEC MS013. 6. N = NUMBER OF PINS.



PACKAGE OUTLINE, .300" SOIC

21-0042 В

/U/IXI/U

### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	8/05	Initial release	_
1	3/09	Changes to remove SO package, style edits	1–7, 12, 13, 14

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